

Markov Model Based Jamming and Anti-Jamming Performance Analysis for Cognitive Radio Networks

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Abstract

In this paper, we conduct a cross-layer analysis of both the jamming capability of the cognitive-radio-based jammers and the anti-jamming capability of the cognitive radio networks (CRN). We use a Markov chain to model the CRN operations in spectrum sensing, channel access and channel switching under jamming. With various jamming models, the jamming probabilities and the throughputs of the CRN are obtained in closed-form expressions. Furthermore, the models and expressions are simplified to determine the minimum and the maximum CRN throughput expressions under jamming, and to optimize important anti-jamming parameters. The results are helpful for the optimal anti-jamming CRN design. The model and the analysis results are verified by simulations.

Keywords

Cognitive Radio, Dynamic Spectrum Access, Jamming, Markov Model, Throughput

1. Introduction

Cognitive radio networks (CRNs) have attracted great attention recently because they can potentially resolve the critical spectrum shortage problem [1]. Under the umbrella of dynamic spectrum access, CRN accesses the spectrum secondarily, *i.e.*, as long as it can guarantee no interference to any primary user (PU) who is using this spectrum at this time in this location [2]. This means that the cognitive radios need to periodically sense the spectrum to detect the PU's activity. They have to vacate the channel immediately whenever the PU activity is detected.

While cognitive radios can realize more flexible spectrum access and higher spectrum efficiency, malicious users can also exploit them to launch more effective attacks, in particular jamming attacks. As a matter of fact, CRN is extremely susceptible to jamming attacks because of its unique requirements in the physical- and MAC-layers, such as the requirement of channel vacating when detecting any PU signals. On the other hand, it is believed that the capability of hopping among many channels gives CRN a unique advantage in improving their anti-jamming performance.

The anti-jamming performance of CRN is a new and interesting research topic that is critical for the secure and reliable CRN design [2]-[4]. Conventionally, anti-jamming study is conducted in the Physical-layer via some anti-jamming modulations, such as spread spectrum, or in the layers above MAC via channel switching. However, even if the CRN has an anti-jam Physical-layer transmission scheme, it may still be sensitive to jamming attacks because of the unique property of CRN in spectrum sensing and spectrum vacating [5]-[7]. In addition, channel switching in CRN is costly considering the required timing/frequency synchronization, channel estimation, handshaking for information exchange and network setup. In particular, the information about the available channels may not be identical among the CRN nodes because of the asynchronous spectrum sensing and the inevitable sensing errors. Extensive handshaking is necessary, which can be extremely timing/bandwidth consuming. Considering the complexity of jamming and anti-jamming interactions in CRN, game theory has also been adopted in anti-jamming research [8]-[10]. Nevertheless, the cost of channel switching has not been addressed sufficiently in these studies.

In this paper, we study the anti-jamming performance of CRN against jammers that are also equipped with similar cognitive radios. We focus on evaluating quantitatively some best jamming parameters as well as some optimal anti-jamming parameters, in particular the effect of number of white space channels. We conduct a cross-layer analysis of both the jamming capability and the anti-jamming capability. To address the CRN specific properties such as channel switching more accurately, we use a Markov chain to model the CRN operations. This provides us an efficient way to analyze the role of channel switching in mitigating jamming attacks. Although Markov model has been widely used for CRN performance analysis [11], its application in anti-jamming study is relatively less. In contrast, we addressed them in [12]-[14], which form the foundation for this paper.

The organization of this paper is as follows. In Section 2, we give the models of the CRN and the cognitive-radio-based jammers. Then in Section 3, we derive the jamming and anti-jamming performance expressions. In Section 4, we analyze and optimize important parameters that are critical for anti-jamming design. Simulations are conducted in Section 5. Conclusions are then given in Section 6.

2. Models of CRN and Jammers

We consider a generic cognitive radio transmission model that includes three states: spectrum sensing, data transmission and channel switching, as illustrated in **Figure 1(a)**. The working sequence of a cognitive radio always begins with the spectrum sensing. If the spectrum sensing indicates that the channel is available for secondary access, then the cognitive radio transmits a data packet, and the model shifts into the data transmission state. If the spectrum sensing indicates the channel is not available, the cognitive radio conducts channel switching, and the model shifts into the channel switching state.

We use the Markov chain in **Figure 1(b)** to model the operation of the CRN, where p_s, p_d, p_c are the probabilities of the CRN in the spectrum sensing, data transmission and channel switching modes, respectively. The transitional probabilities p_{js}, p_{jd}, p_{jc} are the probabilities that the spectrum sensing, data transmission and channel switching procedures are jammed, respectively.

The durations of spectrum sensing slot, data transmission slot, and channel switching slot are T_s, T_d , and T_c . Usually the spectrum sensing duration T_s is much smaller than both T_d and T_c . The CRN has M white space channels to select from. The availability of each channel depends on the activity of the PU and the jammers. The large number of channels is one of the primary advantages of CRN to combat jamming.

We use signal-to-noise-and-interference ratio (SINR) to measure the signal and jamming levels. For the data transmission slot, we assume that the minimum workable SINR is Γ_d . SINR less than Γ_d means that the CRN's data packet transmission is jammed. For the spectrum sensing slot, we assume that if the SINR is larger than the detection threshold Γ_s , then the cognitive radio will make a decision that the channel is occupied by the PU, and is thus not available. Γ_d is usually much larger than Γ_s . We also assume that the minimum SINR for the channel switching procedures is Γ_c , which is usually smaller than Γ_d because the CRN may adopt

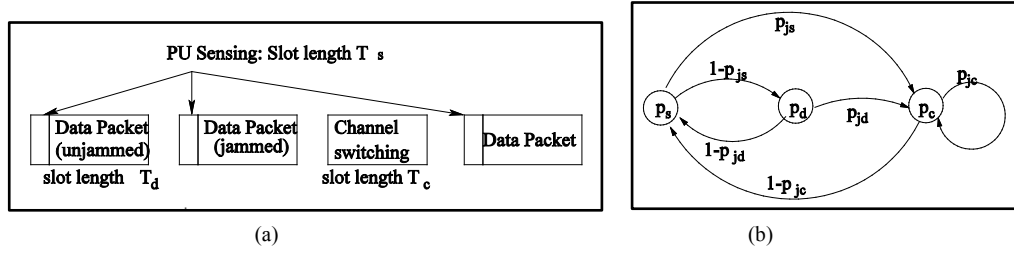


Figure 1. (a) Illustration of cognitive radio transmissions. (b) Markov model for cognitive radio transmissions under jamming.

some more reliable transmission techniques (albeit with lower data rate) such as spread spectrum modulations to increase the reliability of channel switching.

We make the following assumptions about the jammers. 1) There are J jammers. 2) Each jammer uses devices that have similar capabilities as a CRN node, including spectrum sensing and RF transceiving. 3) The jammers do not know the secret keys that the CRN is using for channel selection and communication. Therefore, the jammers do not know which channel the CRN is using. The only way left for the jammers is to randomly select some channels to jam.

Since the jammers can fastly switch among the channels, they may choose to jam multiple channels simultaneously. In this paper the jamming strategies are described by two parameters: the jamming signal duration T_j and the number of channels that are jammed simultaneously. We assume that each jammer has the same transmission power $P_j = P_s$ as the CRN, where P_s is the CRN node's transmission power. The demodulation and signal detection of the CRN receiver depend on the average SINR received during the entire slot. If the jamming duration T_j is smaller, the overall jamming signal power in this slot is lower. But the jammers can jam more channels simultaneously with smaller T_j . We assume that all the jammers adopt the same jamming parameter T_j .

3. CRN Throughput under Multiple Uncooperative Jammers

3.1. Jamming Probabilities

Consider a CRN where a pair of CRN transmitter and receiver is conducting transmission at unit data throughput. A group of J jammers want to jam the CRN transmission so as to reduce the throughput.

First, we consider a data packet transmission slot with slot length T_d and SINR requirement Γ_d . If there are k jamming signals falling in this slot, each with duration T_j , then the SINR is

$$\gamma_d(k) = \frac{P_s \alpha_s^2 T_d}{\sum_{\ell=1}^k P_j \alpha_\ell^2 \min\{T_d, T_j\} + N T_d} \quad (1)$$

where α_s^2 is the power gain of the Rayleigh flat fading channel of the CRN, α_ℓ^2 is the power gain of the Rayleigh flat fading channel of the ℓ th jamming signal, N is the power of the additive white Gaussian noise (AWGN). We assume that α_s^2 and α_ℓ^2 are independent exponential random variables with unit mean. A successful jamming means that $\gamma_d(k) < \Gamma_d$.

The number of jamming signals k is limited to $0 \leq k \leq K_d$, where $K_d \triangleq J \lceil T_d / T_j \rceil$ and $\lceil x \rceil$ denotes the minimum integer that is no less than x . The probability that there are k jamming signals in this slot follows the binomial distribution

$$\mathbb{P}_d[k] = \binom{K_d}{k} p_j^k (1 - p_j)^{K_d - k}, \quad (2)$$

where $p_j = 1/M$ is the probability that a jammer chooses the same channel as the CRN. For simplicity, we do not consider the white space detection errors of the CRN and jammers. White space detection errors may make the available white space channels less for CRN. However, for the jammers, a safer approach might be just to jam every one of the M white space channels.

Proposition 1. If there are k jamming signals with the same jamming duration T_j injected into a data transmission slot of duration T_d , the probability that the data packet transmission is jammed is

$$\mathbb{P}[\gamma_d(k) < \Gamma_d] = 1 - e^{-\frac{N\Gamma_d}{P_s} \left(1 + \frac{P_j \min\{T_d, T_j\}}{P_s T_d} \Gamma_d \right)^{-k}}. \quad (3)$$

Proof. Using (1), we can change $\gamma_d(k) < \Gamma_d$ into $z < \Gamma_d$ with

$$z = \frac{1}{N} P_s \alpha_s^2 - \frac{P_j \min\{T_d, T_j\}}{N T_d} \Gamma_d \sum_{\ell=1}^k \alpha_\ell^2. \quad (4)$$

Since α_s^2 is an exponential random variable with unit mean, its probability density function is $f_{\alpha_s^2}(x) = e^{-x}$, for $x \geq 0$. In addition, $Y = \sum_{\ell=1}^k \alpha_\ell^2$ has Erlong distribution with probability density function $f_Y(y) = y^{k-1} e^{-y} / (k-1)!$, for $y \geq 0$. Due to the independence assumption, their joint distribution is

$$f_{\alpha_s^2, Y}(x, y) = f_{\alpha_s^2}(x) f_Y(y) = \begin{cases} e^{-x} \frac{y^{k-1} e^{-y}}{(k-1)!}, & x \geq 0, y \geq 0 \\ 0, & \text{else} \end{cases}. \quad (5)$$

Then, the probability $\mathbb{P}[\gamma_d(k) < \Gamma_d]$ can be evaluated as

$$\begin{aligned} \mathbb{P}[z < \Gamma_d] &= \iint_{z < \Gamma_d} f_{\alpha_s^2, Y}(x, y) dx dy = \int_0^\infty \frac{y^{k-1} e^{-y}}{(k-1)!} dy \int_0^{\frac{N}{P_s} \Gamma_d \left(1 + \frac{P_j \min\{T_d, T_j\}}{N T_d} y \right)} e^{-x} dx \\ &= \int_0^\infty \frac{y^{k-1} e^{-y}}{(k-1)!} \left[1 - e^{-\frac{N\Gamma_d}{P_s} e^{-\frac{P_j \min\{T_d, T_j\}}{P_s T_d} y}} \right] dy = 1 - \frac{e^{-\frac{N\Gamma_d}{P_s}}}{(k-1)!} \int_0^\infty y^{k-1} e^{-\left(1 + \frac{P_j \min\{T_d, T_j\}}{P_s T_d} \Gamma_d \right) y} dy \end{aligned} \quad (6)$$

The last integration in (6) can be changed to the integration of the Erlong probability density function. According to the property of the Erlong distribution, we can derive (3). ■

Averaging over all possible k , the probability that the data transmission slot is jammed is

$$p_{jd} = \sum_{k=0}^{K_d} \mathbb{P}[\gamma_d(k) < \Gamma_d] \mathbb{P}_d[k], \quad (7)$$

which can be evaluated by using (2) and (3).

Proposition 2. For the channel switching slot with duration T_c and required SINR Γ_c , the probability of being jammed is

$$p_{jc} = \sum_{k=0}^{K_c} \mathbb{P}[\gamma_c(k) < \Gamma_c] \mathbb{P}_c[k], \quad (8)$$

where the maximum number of jamming signals in this slot is $K_c = J \lceil T_c / T_j \rceil$, the jamming probability under k jamming signals is

$$\mathbb{P}[\gamma_c(k) < \Gamma_c] = 1 - e^{-\frac{N\Gamma_c}{P_s} \left(1 + \frac{P_j \min\{T_c, T_j\}}{P_s T_c} \Gamma_c \right)^{-k}}, \quad (9)$$

and the probability of having k jamming signals is

$$\mathbb{P}_c[k] = \binom{K_c}{k} p_j^k (1 - p_j)^{K_c - k}. \quad (10)$$

Proof. We can derive (8)-(10) by following the proof of the Proposition 1, and by replacing T_d and Γ_d with T_c and Γ_c , respectively. ■

The spectrum sensing slot is different from either the data packet slot or the channel switching slot be-

cause the SINR $\gamma_s(k)$ is in fact the interference (jamming) to noise ratio

$$\gamma_s(k) = \frac{\sum_{\ell=1}^k P_j \min\{T_s, T_j\} \alpha_\ell^2}{NT_s}. \quad (11)$$

Usually the CRN is highly sensitive in PU sensing, which means that there is an extremely small sensing threshold Γ_s . By making $\gamma_s(k) \geq \Gamma_s$, the jammers disguise the PUs to force the CRN to conduct channel switching, which defines the jamming of the sensing slots.

Proposition 3. For the channel sensing slot with duration T_s and sensing threshold Γ_s , the probability of being jammed is

$$p_{js} = \sum_{k=0}^{K_s} \mathbb{P}[\gamma_s(k) \geq \Gamma_s] \mathbb{P}_s[k], \quad (12)$$

where the maximum number of jamming signals in this slot is $K_s = J\lceil T_s/T_j \rceil$, the probability of having k jamming signals is

$$\mathbb{P}_s[k] = \binom{K_s}{k} p_j^k (1-p_j)^{K_s-k}, \quad (13)$$

and the jamming probability under k jamming signals is

$$\mathbb{P}[\gamma_s(k) \geq \Gamma_s] = 1 - \frac{\gamma(k, a)}{(k-1)!} = \sum_{n=0}^{k-1} \frac{1}{n!} e^{-a} a^n, \quad (14)$$

where $a = \frac{NT_s\Gamma_s}{P_j \min\{T_s, T_j\}}$ and $\gamma(k, a)$ is the lower incomplete Gamma function.

Proof. The equation (13) can be derived similarly as $\mathbb{P}_d[k]$ in (2) by replacing T_d with T_s . To derive (14), from the SINR definition (11) and the Erlong distribution of $Y = \sum_{\ell=1}^k \alpha_\ell^2$, we have

$$\mathbb{P}[\gamma_s(k) \geq \Gamma_s] = \mathbb{P}\left[Y \geq \frac{NT_s\Gamma_s}{P_j \min\{T_s, T_j\}}\right] = 1 - \int_0^{\frac{NT_s\Gamma_s}{P_j \min\{T_s, T_j\}}} f_Y(y) dy. \quad (15)$$

From the property of the Erlong distribution, the integration of (15) leads to (14). ■

3.2. Throughput of CRN under Jamming

With the jamming probabilities p_{jd}, p_{jc}, p_{js} derived in Equations (7), (8) and (12), we can calculate the steady state probabilities of the three states p_s, p_d and p_c of the Markov model in **Figure 1(b)** by solving the equation

$$\begin{bmatrix} -1 & 1-p_{jd} & 1-p_{jc} \\ 1-p_{js} & -1 & 0 \\ p_{js} & p_{jd} & p_{jc}-1 \end{bmatrix} \begin{bmatrix} p_s \\ p_d \\ p_c \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}. \quad (16)$$

We need the constraint $p_s + p_d + p_c = 1$ for (16) to have a unique solution.

From (16) we can find that the system stays in the data packet transmission slots with probability

$$p_d = \frac{(1-p_{js})(1-p_{jc})}{2-p_{jc} + (p_{jd}-p_{jc})(1-p_{js})}. \quad (17)$$

However, some of the data packets are lost due to jamming. Considering the jamming probability p_{jd} , the data packet transmission is successful with probability $p_d(1-p_{jd})$.

We define the normalized average throughput of the CRN transmission as

$$R = \frac{p_d(1-p_{jd})T_d}{p_sT_s + p_dT_d + p_cT_c}. \quad (18)$$

Proposition 4. Considering the throughput definition (18), the throughput of CRN under jamming is

$$R = \frac{(1-p_{js})(1-p_{jc})(1-p_{jd})T_d}{(1-p_{jc})T_s + (1-p_{js})(1-p_{jc})T_d + (p_{jd} + p_{js} - p_{jd}p_{js})T_c}. \quad (19)$$

Proof. From (16), we can describe p_d and p_c by p_s as

$$p_d = (1-p_{js})p_s, \quad p_c = \frac{p_{jd} + p_{js} - p_{jd}p_{js}}{1-p_{jc}}p_s. \quad (20)$$

Substituting (20) into (18), we can get (19). ■

4. Analysis of Anti-Jamming Performance

4.1. A Closed-Form Throughput Expression

Although (19) can be used to evaluate numerically the anti-jamming performance, further analytical study is difficult. In this section, we first adopt some reasonable simplifications to simplify the jamming probability and the throughput expressions. Then, we analyze the anti-jamming performance by deriving the maximum and minimum throughputs under jamming.

For the jammer model, instead of considering J non-cooperative jammers that randomly inject k jamming signals with duration T_j and power P_j , we assume in this section that they cooperatively inject a single jamming signal of duration T_j (which equals to the CRN slot lengths without loss of generality). The total power is $P_J = JP_j$. We assume that the jammers can jam K channels simultaneously, and the jamming power sent to each channel is $P_K = P_J/K$. We define $P_0 = 0$. We also assume that the maximum number of channels that these jammers can jam simultaneously is K_J , which means $0 \leq K \leq K_J$.

For the data transmission slot, if there is jamming signal, the cognitive radio's received signal's SINR is $\gamma_d = P_s\alpha_s^2 / (P_K\alpha_j^2 + N)$. If there is no jamming signal, the SINR becomes $\gamma_{d'} = P_s\alpha_s^2 / N$.

Since there are M white space channels, if the jammers randomly select K channels to jam, then they have probability K/M of sending a jamming signal to the channel being used by the CRN. Therefore, the probability that the data transmission is jammed (including the case that the channel gain is too small to transmit data successfully) can be written as

$$p_{jd} = \mathbb{P}[\gamma_d < \Gamma_d] \frac{K}{M} + \mathbb{P}[\gamma_{d'} < \Gamma_d] \left(1 - \frac{K}{M}\right). \quad (21)$$

Note that $\mathbb{P}[\gamma_d < \Gamma_d] = \mathbb{P}[P_s\alpha_s^2 - P_K\Gamma_d\alpha_j^2 < N\Gamma_d]$. Similar to the proof of the Proposition 1, we can find

$$\mathbb{P}[\gamma_d < \Gamma_d] = 1 - e^{-\frac{N\Gamma_d}{P_s}} \frac{P_s}{P_s + P_K\Gamma_d}, \quad \mathbb{P}[\gamma_{d'} < \Gamma_d] = 1 - e^{-\frac{N\Gamma_d}{P_s}}. \quad (22)$$

Therefore, the probability that the data transmission is jammed can be derived from (21)-(22) as

$$p_{jd} = 1 - e^{-\frac{N\Gamma_d}{P_s}} \left(1 + \frac{P_J\Gamma_d}{M(P_s + P_K\Gamma_d)}\right). \quad (23)$$

Next, we consider the channel sensing slots. In case of absence of PU, if there is jamming signal in this sensing slot, then we have SINR $\gamma_s = P_K\alpha_j^2 / N$. Otherwise, the SINR becomes simply $1/N$. For jamming probability, we just need to consider γ_s . The probability of having jamming signal in this sensing slot is similarly K/M . Therefore, the probability that the sensing slot is jammed can be found as

$$p_{js} = \mathbb{P}[\gamma_s \geq \Gamma_s] \frac{K}{M} = \frac{P_J}{MP_K} e^{-\frac{N\Gamma_s}{P_K}} \quad (24)$$

because γ_s has exponential distribution.

Finally, because channel switching is usually more jamming-resistant than channel sensing and data transmission, we let $p_{jc} = 0$. Furthermore, without loss of generality, we let $T_c = T_d$, which means that the CRN waits for a full data slot before switching to a new channel. Then the throughput (19) can be readily deduced into a close-form expression

$$R = \frac{e^{-\frac{N\Gamma_d}{P_s}}}{1 + \frac{T_s}{T_d}} \left(1 + \frac{\Gamma_d P_J}{MP_s + MP_K \Gamma_d} \right) \left(1 - e^{-\frac{N\Gamma_s}{P_K} \frac{P_J}{MP_K}} \right). \quad (25)$$

4.2. Anti-Jamming Throughput Analysis

One of the major parameters for the jammers to adjust jamming attacks is the jamming signal strength P_K , or equivalently, the number of channels to jam simultaneously $K = P_J/P_K$. In contrast, one of the major parameters for the CRN to mitigate jamming is the number of white space channels M .

If considering just P_K and M , the optimal anti-jamming performance of CRN can be found from the max-min optimization $\max_{M>0} \min_{0 \leq P_K \leq P_J} R$.

First, let us analyze the jammer's best strategy to minimize the CRN throughput. Define $y = P_K/P_J = 1/K$, and rewrite the throughput (25) into

$$R(y) = \frac{e^{-\frac{N\Gamma_d}{P_s}}}{1 + \frac{T_s}{T_d}} \left(1 + \frac{\Gamma_d}{\frac{P_s}{P_J} + y\Gamma_d} \right) \left(1 - e^{-\frac{N\Gamma_s}{P_J y} \frac{1}{My}} \right). \quad (26)$$

Note that the range of y is $1/K_J \leq y \leq 1$. If y is extremely small, the item $e^{-N\Gamma_s/P_J y}/My \approx 0$ can be omitted from $R(y)$. In this case, the derivative

$$\frac{\partial R(y)}{\partial y} \approx -\frac{e^{-\frac{N\Gamma_d}{P_s}}}{1 + \frac{T_s}{T_d}} \frac{\Gamma_d^2}{M \left(\frac{P_s}{P_J} + \Gamma_d y \right)^2} < 0, \quad (27)$$

which means that $R(y)$ is a monotone decreasing function of y when y is extremely small.

When y is not so small, because $N\Gamma_s/P_J$ is usually a very small number, we let $e^{-N\Gamma_s/P_J y} \approx 0$. In this case, by taking the derivative of (26) with respect to y , we can easily find that

$$\frac{\partial R(y)}{\partial y} \approx \frac{e^{-\frac{N\Gamma_d}{P_s}}}{1 + \frac{T_s}{T_d}} \frac{1}{M} \left(\frac{1}{\frac{P_s}{P_J \Gamma_d} + y} + \frac{1}{y} \right) > 0, \quad (28)$$

which means that the throughput $R(y)$ becomes a monotone increasing function for relatively large y . Therefore, the minimum $R(y)$ should happen with some extremely small y values, whereas the maximum throughput happens either when $y = 1$ or $y = 1/K_J$. The former means that the jammers just jam one channel at a time, while the latter means the weakest jamming signal is used. The maximum throughput is thus $R_{\max} = \max \{R(1), R(1/K_J)\}$, which can be calculated from (26).

The optimal jamming parameter y for the jammers to minimize the CRN throughput is shown below.

Proposition 5. For the jammers, the (approximately) optimal jamming parameter is

$$y_o = \max \left\{ \frac{1}{K_J}, \frac{N\Gamma_s}{P_J} \right\} \quad (29)$$

which minimizes the CRN throughput into $R_{\min} = R(y_o)$.

Proof. From (26), we can take the derivative $\partial R(y)/\partial y$ and let it be zero to find the optimal y . After some straightforward deductions, we can get

$$\left(1 - \frac{1}{My} e^{-\frac{N\Gamma_s}{P_j y}}\right) \frac{\Gamma_d^2}{M(P_s/P_j + \Gamma_d y)^2} - \left(1 + \frac{\Gamma_d/M}{P_s/P_j + \Gamma_d y}\right) e^{-\frac{N\Gamma_s}{P_j y}} \frac{1}{My^2} \left(1 - \frac{N\Gamma_s}{P_j y}\right) = 0. \quad (30)$$

Unfortunately, (30) is too complex to find closed-form solutions to y . Therefore, as an approximation, we consider the major items only. Because the minimum $R(y)$ happens when y is extremely small, we can consider only those items in (30) involving $O(y^{-2})$ and $O(y^{-3})$. Then (30) can be approximately simplified to

$$\left(1 + \frac{\Gamma_d/M}{P_s/P_j + \Gamma_d y}\right) e^{-\frac{N\Gamma_s}{P_j y}} \frac{1}{My^2} \left(1 - \frac{N\Gamma_s}{P_j y}\right) = 0 \quad (31)$$

which gives solution $y = N\Gamma_s/P_j$. Considering the practical range of y and the monotone property of $R(y)$, we can derive (29). The throughput can be obtained by applying y_o into (26). ■

If M is large and $y_o = N\Gamma_s/P_j$, then the minimum throughput becomes

$$R_{\min} = \frac{e^{-\frac{N\Gamma_d}{P_s}}}{1 + \frac{T_s}{T_d}} \left(1 + \frac{P_j/M}{P_j \Gamma_d + N\Gamma_s}\right) \left(1 - \frac{e^{-1} P_j}{MN\Gamma_s}\right) \leq \frac{e^{-\frac{N\Gamma_d}{P_s}}}{1 + \frac{T_s}{T_d}} \left(1 - \frac{P_j^2}{M^2 N^2 \Gamma_s^2}\right). \quad (32)$$

On the other hand, if M is small so the jammer can jam all the channels with $y = 1/M$, then the throughput becomes

$$R_{\min} = \frac{e^{-\frac{N\Gamma_d}{P_s}}}{1 + \frac{T_s}{T_d}} \left(1 + \frac{\Gamma_d}{M \frac{P_s}{P_j} + \Gamma_d}\right) \left(1 - e^{-\frac{MN\Gamma_s}{P_j}}\right). \quad (33)$$

From (32)-(33), it can be seen that M should be extremely large (e.g., several hundreds) for moderately high throughput. The CRN can increase M to mitigate jamming. Unfortunately, from (32) we can see that the throughput increases according to $O(1-1/M^2)$ only, which means larger M only brings smaller throughput increase, or the throughput increase tends to saturate at large M . Alternatively, the CRN may reduce the length of spectrum sensing slot T_s or increase the spectrum sensing threshold Γ_s to increase the throughput. But this may increase interference to PUs. Therefore, anti-jamming CRN design is a challenging issue.

5. Simulations

In this section, we use simulations to verify the analysis results derived in Sections 3 and 4. Specifically, the normalized average throughput R and the probability of transmitting unjammed data packets $p_d(1-p_{jd})$ were evaluated. The following parameters were used: $M=100$, $J=10$, $T_d=5$, $T_c=10$, $T_s=0.25$, $\Gamma_d=15$ dB, $\Gamma_c=10$ dB, $\Gamma_s=-15$ dB, $P_s=P_j=-80$ dBm, and $N=-100$ dBm.

First, we verify the results in Section 3. For the jammers, we tested the jamming signals with duration T_j from 0 (jamming-free) up to the duration of T_c since T_c was the longest slot length. For the theoretical results, we used (19) to calculate R and used (7) and (17) to calculate $p_d(1-p_{jd})$.

The simulation results are shown in **Figure 2**. From **Figure 2(a)**, we can see that the theoretical results fit well with the simulated results, which demonstrated the validity of the modelling and analysis. Compared to the jamming-free throughput ($T_j/T_c=0$) which was near unity, the throughput drastically reduced to just below 0.3 when facing 10 jammers that used small T_j . Even with 100 channels to hop from, the CRN throughput still suffered from detrimental effect of jamming.

In **Figure 2(b)**, we evaluated the anti-jamming performance of CRN when the CRN could hop among a large number of white space channels. It can be seen that while increasing the channel number M could drastically increase the anti-jamming capability of CRN, such a benefit tended to saturate after tens of channels had been used. Even with 1000 white space channels, the average throughput were still just around 0.6. In contrast, the jammers could reduce this benefit by just using a few more jammers.

Next, we use simulations to verify that the analysis results of Section 4. For the jammer, we evaluated the jamming parameters y from 0 (jamming-free) up to 0.1. The simulation results are shown in **Figure 3**. From the

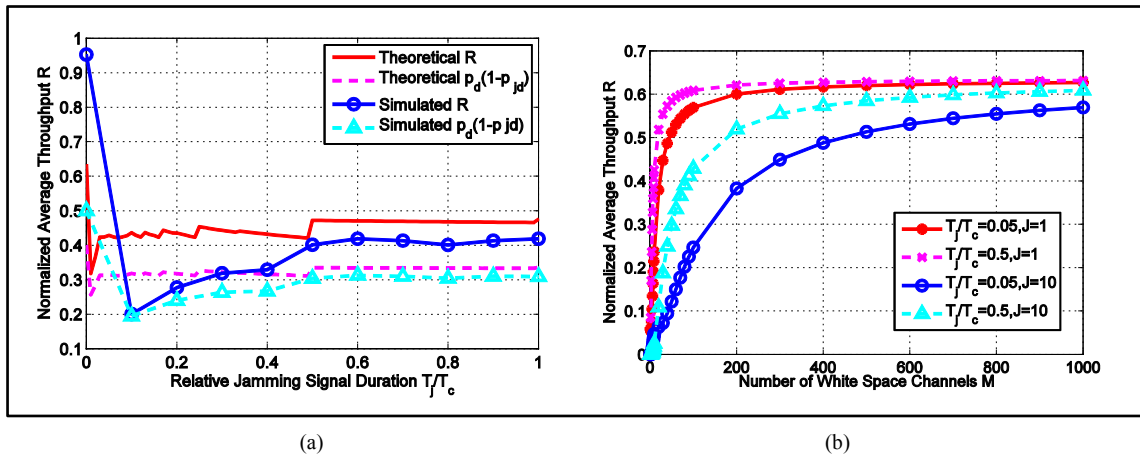


Figure 2. (a) Average throughput and probability of transmitting unjammed data packets under various jamming parameters. (b) Average throughput as function of number of white space channels, under various jamming parameters.

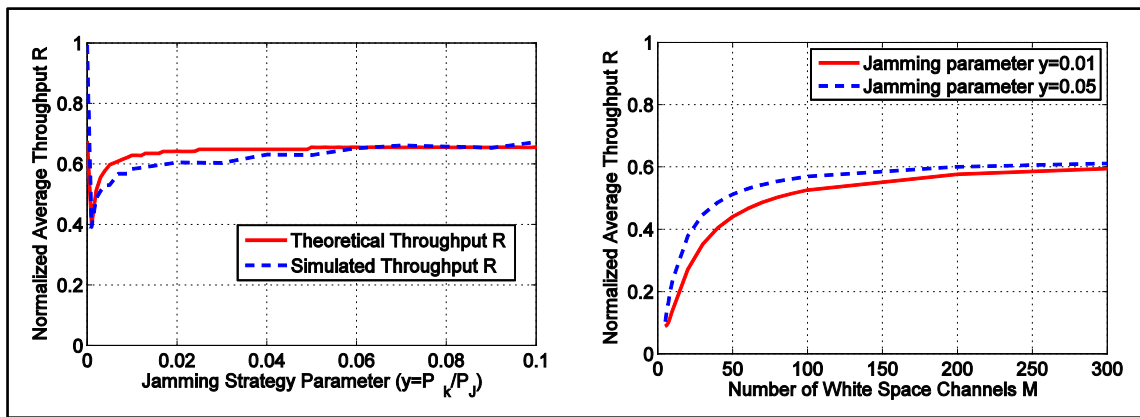


Figure 3. Comparison of simulation results to the analysis results of the average throughput.

results, we can see that the analysis results fit well with the simulated results, which demonstrated the validity of the analysis. It clearly showed that the throughput reduced with γ when γ was extremely small, but increased with γ when γ became larger.

6. Conclusion

In this paper, with a Markov model of the cognitive radio transmissions, both the jamming performance of the cognitive-radio-based jammers and the anti-jamming performance of the CRN are analyzed. Expressions of the CRN average throughput and jamming probabilities are derived. Some optimal jamming parameters and anti-jamming parameters are analyzed, in particular the number of white space channels, which are verified by simulations. The results indicate that the CRN is extremely susceptible to jamming attacks, and it suffers from a saturation effect when combating jamming attacks by increasing the number of white space channels.

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